Passivity-Based Gain-Scheduled Control with Scheduling Matrices

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Motivating Example

Consider the control of the rigid two-link robotic manipulator in Figure 1.

For simplicity, the robot is assumed to be:

- 1. Planar.
- 2. Fixed at the base.

The equations of motion of this two-link robot are given by

$$\mathbf{M}(\mathbf{q}(t))\ddot{\mathbf{q}}(t) = \mathbf{f}_{\mathrm{non}}(\mathbf{q}(t), \dot{\mathbf{q}}(t)) + \mathbf{u}(t),$$

where

- $\mathbf{M}(\mathbf{q}(t)) = \mathbf{M}^{\mathsf{T}}(\mathbf{q}(t)) \succ 0$ is the mass matrix,
- $\mathbf{f}_{non}(\mathbf{q}(t), \dot{\mathbf{q}}(t))$ captures the nonlinear inertial and Coriolis forces,
- $\mathbf{u}(t) = \begin{bmatrix} \tau_1(t) & \tau_2(t) \end{bmatrix}^T$ are the joint torques, and
- $\mathbf{q}(t) = \begin{bmatrix} \theta_1(t) & \theta_2(t) \end{bmatrix}^{\mathsf{T}}$ are the generalized coordinates.

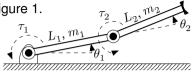


Figure 1: Two-link robotic manipulator.

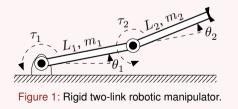
Control Objective

The control objective is to have the two-link robot track the position and rate trajectory in Figure 2.

Nonlinear System

$$\mathbf{M}(\mathbf{q}(t))\ddot{\mathbf{q}}(t) = \mathbf{f}_{\mathrm{non}}(\mathbf{q}(t), \dot{\mathbf{q}}(t)) + \mathbf{u}(t).$$

The mapping between the joint torques to joint rates, $\mathbf{u}(t) \rightarrow \dot{\mathbf{q}}(t)$, is **passive**.



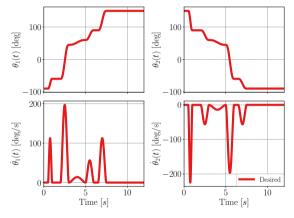


Figure 2: Desired angle position and rate.

What is Passivity?

Definition (Passivity [Márquez, 2003])

Consider a square system with input $\mathbf{u} \in \mathcal{L}_{2e}$ and output $\mathbf{y} \in \mathcal{L}_{2e}$ mapped through the operator $\mathcal{G} : \mathcal{L}_{2e} \to \mathcal{L}_{2e}$. The system \mathcal{G} is

b passive if $\exists \beta \in \mathbb{R}_{\leq 0}$ s.t.

$$\langle \mathbf{y}, \mathbf{u} \rangle_T \ge \beta, \qquad \forall \mathbf{u} \in \mathcal{L}_{2e}, \quad \forall T \in \mathbb{R}_{\ge 0},$$

• very strictly passive (VSP) if $\exists \delta, \varepsilon \in \mathbb{R}_{>0}$ and $\exists \beta \in \mathbb{R}_{\leq 0}$ s.t.

 $\langle \mathbf{y}, \mathbf{u} \rangle_T \ge \beta + \delta \|\mathbf{u}\|_{2T}^2 + \varepsilon \|\mathbf{y}\|_{2T}^2, \quad \forall \mathbf{u} \in \mathcal{L}_{2e}, \quad \forall T \in \mathbb{R}_{\ge 0},$

- input strictly passive (ISP) if $\delta \in \mathbb{R}_{>0}$ and $\varepsilon = 0$,
- output strictly passive (OSP) if $\varepsilon \in \mathbb{R}_{>0}$ and $\delta = 0$.

Why Should You Care?

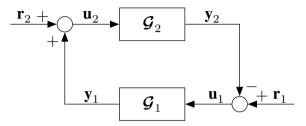


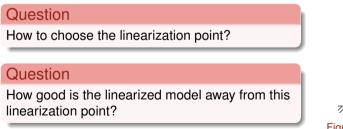
Figure 3: The negative feedback interconnection of two systems \mathcal{G}_1 and \mathcal{G}_2 .

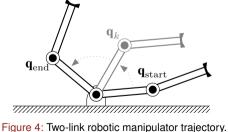
Theorem (Passivity Theorem [Márquez, 2003])

Consider the negative feedback interconnection of $\mathcal{G}_1 : \mathcal{L}_{2e} \to \mathcal{L}_{2e}$ and $\mathcal{G}_2 : \mathcal{L}_{2e} \to \mathcal{L}_{2e}$ in Figure 3. Provided \mathcal{G}_1 is **passive** and \mathcal{G}_2 is **VSP**, the negative feedback interconnection is \mathcal{L}_2 -stable.

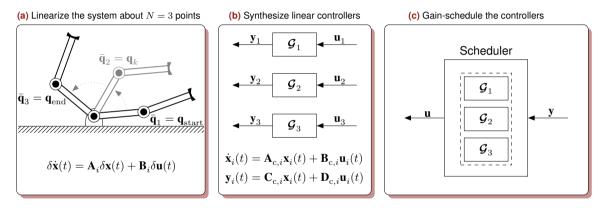
VSP Controller Synthesis

- Strictly positive real (SPR) controllers with feedthrough are in turn VSP [Márquez, 2003].
- The Kalman-Yakubovich-Popov (KYP) lemma and gain matrix from the linear-quadratic regulator (LQR) problem can be used to synthesize the SPR controllers [Benhabib et al., 1981].
- LQR problem needs the linearized model of the system dynamics about a linearization point.

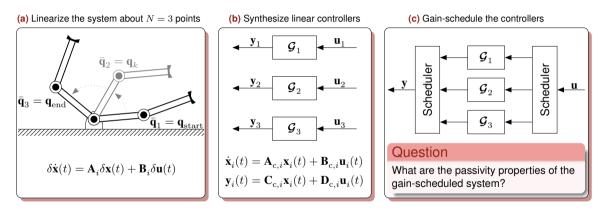




Gain-Scheduling Linear VSP Controllers



Gain-Scheduling Linear VSP Controllers



Passivity Preserving Gain-Scheduling Architecture

Within the context of **passivity**-based control, the gain-scheduling architecture in Figure 5 was proposed in [Damaren, 1996].

Scheduling is achieved via the scalar scheduling function $s_i(\zeta(t), \mathbf{x}(t), t)$, where

- $\mathbf{x}(t)$ is the state of the plant under control,
- ζ(t) represent any external signal convenient for scheduling.

The gain-scheduled controller is

- passive if the subcontrollers are passive,
- ISP if the subcontrollers are ISP.

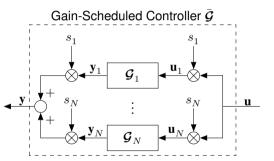
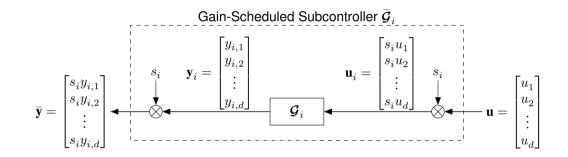


Figure 5: Gain-scheduling architecture [Damaren, 1996]. The node \otimes performs scalar-vector multiplication using the scalar scheduling signal $s_i(\zeta(t), \mathbf{x}(t), t)$.

Passivity Preserving Gain-Scheduling Literature

- Gain-Scheduled SPR Controllers for Nonlinear Flexible Systems [Damaren, 1996]
- Design of Gain-Scheduled Strictly Positive Real Controllers Using Numerical Optimization for Flexible Robotic Systems [Forbes, Damaren, 2010]
- A Very Strictly Passive Gain-Scheduled Controller: Theory and Experiments [Walsh, Forbes, 2016]
- Very Strictly Passive Controller Synthesis With Affine Parameter Dependence [Walsh, Forbes, 2018]
- Gain-Scheduled Control for an Antenna with Multiple Collocated Sensors and Actuators [Lang, Damaren, 2018]

Scalar-Gain-Scheduling Architecture

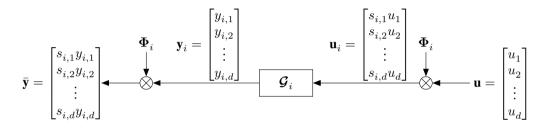


Question

• Why should $(u_1, y_{i,1}), (u_2, y_{i,2}), \dots, (u_d, y_{i,d})$ all be scheduled by the same s_i ?

Gain-Scheduling Goal

Consider the case where the scheduling function is a **diagonal matrix**, $\Phi_i = \text{diag}(s_{i,1}, \dots, s_{i,d})$:



Questions

- Can the scheduling function be any matrix (not necessarily diagonal)?
- Can the input scheduling function be different from the output scheduling function?

Novel Matrix-Gain-Scheduling Architecture

- Consider subcontrollers $\mathcal{G}_1, \mathcal{G}_2, \dots, \mathcal{G}_N$ of the form $\mathbf{y}_i(t) = (\mathcal{G}_i \mathbf{u}_i)(t)$, for $i \in \mathcal{N} = \{1, \dots, N\}$.
- ► The subcontrollers could be **linear** or **nonlinear**.
- The input-output map of $\overline{\mathcal{G}}$ is given by

$$\begin{split} \mathbf{u}_i(t) &= \mathbf{\Phi}_i(\boldsymbol{\zeta}(t), \mathbf{x}(t), t) \mathbf{u}_c(t), \\ \mathbf{y}_c(t) &= \sum_{i \in \mathcal{N}} \alpha_i \mathbf{\Phi}_i^{\mathsf{T}}(\boldsymbol{\zeta}(t), \mathbf{x}(t), t) \mathbf{y}_i(t), \end{split}$$

for
$$\mathbf{u}_{\mathrm{c}}, \mathbf{y}_{\mathrm{c}} \in \mathbb{R}^{n}$$
, $\alpha_{i} \in \mathbb{R}_{>0}$, and $\Phi_{i} \in \mathbb{R}^{n \times n}$ for all $i \in \mathcal{N}$.

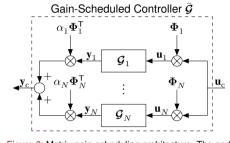


Figure 6: Matrix-gain-scheduling architecture. The node \otimes performs matrix-vector multiplication using the scheduling matrices $\Phi_i(\boldsymbol{\zeta}(t), \mathbf{x}(t), t)$. The positive constants α_i are used to scale the gain.

Question

Knowing the subcontrollers, \mathcal{G}_i , have certain passivity properties, what can be said about the **passivity properties of the matrix-gain-scheduled controller**, $\bar{\mathcal{G}}$?

(1a) (1b)

Scheduling Matrix Properties

Definition (Active Scheduling Matrices)

The scheduling matrices are said to be:

- Active if at all times, there exists at least one nonzero scheduling matrix.
- Strongly active if at all times, there exists at least one full rank scheduling matrix.

Lemma

Provided the scheduling matrices are strongly active, then

$$\sum_{i \in \mathcal{N}} \lambda_{\min} \Big(\boldsymbol{\Phi}_i^\mathsf{T}(t) \boldsymbol{\Phi}_i(t) \Big) = \sum_{i \in \mathcal{N}} \nu_i^2(t) > 0, \quad \forall t \in [0, T],$$

where $\nu_i(t)$ is the smallest singular value of $\Phi_i(t)$.

Why have Φ_i and Φ_i^T ?

Review (Input-Output Map of $\overline{\mathcal{G}}$)

The input-output map of $\bar{\mathcal{G}}$ is given by

$$\begin{split} \mathbf{u}_i(t) &= \mathbf{\Phi}_i(t) \mathbf{u}_c(t), \\ \mathbf{y}_c(t) &= \sum_{i \in \mathcal{N}} \alpha_i \mathbf{\Phi}_i^\mathsf{T}(t) \mathbf{y}_i(t). \end{split}$$

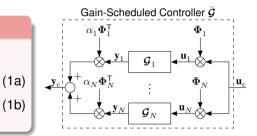


Figure 6: Matrix-gain-scheduling architecture.

Remark

Using the input-output map of the gain-scheduled controller $\bar{\mathcal{G}}$ in (1), it follows that

$$\langle \mathbf{u}_{\mathrm{c}}, \mathbf{y}_{\mathrm{c}} \rangle_{T} = \sum_{i \in \mathcal{N}} \int_{0}^{T} \alpha_{i} \mathbf{u}_{\mathrm{c}}^{\mathsf{T}}(t) \mathbf{\Phi}_{i}^{\mathsf{T}}(t) \mathbf{y}_{i}(t) \, \mathrm{d}t = \sum_{i \in \mathcal{N}} \int_{0}^{T} \alpha_{i} (\mathbf{\Phi}_{i}(t) \mathbf{u}_{\mathrm{c}}(t))^{\mathsf{T}} \mathbf{y}_{i}(t) \, \mathrm{d}t = \sum_{i \in \mathcal{N}} \alpha_{i} \, \langle \mathbf{u}_{i}, \mathbf{y}_{i} \rangle_{T} \, .$$

Main Contribution

Theorem (Gain-Scheduling of ISP subcontrollers)

Consider N ISP subcontrollers $\mathcal{G}_1, \mathcal{G}_2, \dots, \mathcal{G}_N$ of the form $\mathbf{y}_i(t) = (\mathcal{G}_i \mathbf{u}_i)(t)$ satisfying

$$\langle \mathbf{u}_i, \mathbf{y}_i \rangle_T \geq \beta_i + \delta_i \|\mathbf{u}_i\|_{2T}^2, \qquad \forall \mathbf{u}_i \in \mathcal{L}_{2e}, \quad \forall T \in \mathbb{R}_{\geq 0},$$

for $i \in \mathcal{N} = \{1, \dots, N\}$ with $\beta_i \in \mathbb{R}_{\leq 0}$ and $\delta_i \in \mathbb{R}_{>0}$.

Provided the scheduling matrices are **strongly active**, the gain-scheduled controller $\overline{\mathcal{G}}$ is ISP with

$$\langle \mathbf{u}_{\mathrm{c}}, \mathbf{y}_{\mathrm{c}}
angle_T \geq \hat{eta} + \hat{\delta} \|\mathbf{u}_{\mathrm{c}}\|_{2T}^2, \qquad orall \mathbf{u}_{\mathrm{c}} \in \mathcal{L}_{2e}, \quad orall T \in \mathbb{R}_{\geq 0},$$

where

$$\hat{\beta} = \sum_{i \in \mathcal{N}} \alpha_i \beta_i \leq 0, \qquad \hat{\delta} = \delta_{\min} \nu_{\inf} > 0, \qquad \delta_{\min} = \min_{i \in \mathcal{N}} \alpha_i \delta_i > 0, \qquad \nu_{\inf} = \inf_{t \in [0,T]} \sum_{i \in \mathcal{N}} \nu_i^2(t) > 0.$$

Main Contribution

Theorem (Gain-Scheduling of OSP subcontrollers)

Consider N OSP subcontrollers ${\cal G}_1, {\cal G}_2, \ldots, {\cal G}_N$ of the form ${\bf y}_i(t) = ({\cal G}_i {\bf u}_i)(t)$ satisfying

$$\langle \mathbf{u}_i, \mathbf{y}_i \rangle_T \geq \beta_i + \varepsilon_i \|\mathbf{y}_i\|_{2T}^2, \qquad \forall \mathbf{u}_i \in \mathcal{L}_{2e}, \quad \forall T \in \mathbb{R}_{\geq 0},$$

for $i \in \mathcal{N} = \{1, \dots, N\}$ with $\beta_i \in \mathbb{R}_{\leq 0}$ and $\varepsilon_i \in \mathbb{R}_{>0}$.

Provided the scheduling matrices are **active**, the gain-scheduled controller $\overline{\mathcal{G}}$ is OSP with

$$\langle \mathbf{u}_{c}, \mathbf{y}_{c} \rangle_{T} \geq \bar{\beta} + \bar{\varepsilon} \| \mathbf{y}_{c} \|_{2T}^{2}, \qquad \forall \mathbf{u}_{c} \in \mathcal{L}_{2e}, \quad \forall T \in \mathbb{R}_{\geq 0},$$

where

$$\bar{\beta} = \sum_{i \in \mathcal{N}} \alpha_i \beta_i \leq 0, \qquad \bar{\varepsilon} = \frac{\varepsilon_{\min}}{\alpha_{\max}^2 \bar{\sigma}_{\boldsymbol{\Psi}}^2} > 0, \qquad \varepsilon_{\min} = \min_{i \in \mathcal{N}} \alpha_i \varepsilon_i > 0, \qquad \bar{\sigma}_{\boldsymbol{\Psi}} = \sup_{t \in [0,T]} \sigma_{\boldsymbol{\Psi}}(t) > 0,$$

and $\sigma_{\Psi}(t)$ is the largest singular value of $\Psi(t) = \begin{bmatrix} \Phi_1(t)^{\mathsf{T}} & \dots & \Phi_N(t)^{\mathsf{T}} \end{bmatrix}$.

Discussion

Review (Theorem 1: ISP Case)

 $\bar{\mathcal{G}}$ is ISP if each \mathcal{G}_i is ISP and the scheduling matrices are **strongly active**.

Review (Theorem 2: OSP Case)

 $\bar{\mathcal{G}}$ is OSP if each \mathcal{G}_i is OSP and the scheduling matrices are **active**.

Remark (VSP Case)

Consider N VSP subcontrollers $\mathcal{G}_1, \mathcal{G}_2, \ldots, \mathcal{G}_N$.

- Each \mathcal{G}_i is also **ISP and OSP simultaneously**.
- The condition required for $\overline{\mathcal{G}}$ to be ISP is **more restrictive** than that for OSP.

Therefore, the matrix-gain-scheduling of N VSP subcontrollers satisfies Theorem 1 and Theorem 2 simultaneously, provided the scheduling matrices are strongly active.

Comparison with Existing Literature

(a) Gain-scheduling ISP systems [Damaren, 1996] and [Forbes, Damaren, 2010]

The scheduling signals, $s_i(t)$, are assumed to satisfy:

1.
$$\sum_{i \in \mathcal{N}} s_i^2(t) > 0$$
,

2.
$$s_i \in \mathcal{L}_{2e} \cap \mathcal{L}_{\infty}$$
.

The gain-scheduled controller is ISP with coefficient

$$\hat{\delta} = \inf_{t \in [0,T]} \sum_{i \in \mathcal{N}} s_i^2(t) \delta_{\min},$$

with $\delta_{\min} = \min_{i \in \mathcal{N}} \delta_i$.

(b) Matrix-gain-scheduling ISP systems

The scheduling matrices, $\mathbf{\Phi}_i(t)$ are assumed to satisfy:

- 1. Strongly active: $\sum_{i \in \mathcal{N}} \nu_i^2(t) > 0$,
- **2.** Bounded: $\sup_{t \in [0,T]} \| \Phi_i(t) \|_2^2 < \infty$.

The gain-scheduled controller is ISP with coefficient

$$\hat{\delta} = \inf_{t \in [0,T]} \sum_{i \in \mathcal{N}} \nu_i^2(t) \delta_{\min},$$

with
$$\delta_{\min} = \min_{i \in \mathcal{N}} \alpha_i \delta_i$$
.

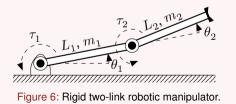
Example: Control of a Two-Link Robotic Manipulator

The control objective is to have the two-link robot track the position and rate trajectory in Figure 2.

Nonlinear System

$$\mathbf{M}(\mathbf{q}(t))\ddot{\mathbf{q}}(t) = \mathbf{f}_{\mathrm{non}}(\mathbf{q}(t), \dot{\mathbf{q}}(t)) + \mathbf{u}(t).$$

The mapping between the joint torques to joint rates, $\mathbf{u}(t) \rightarrow \dot{\mathbf{q}}(t)$, is **passive**.



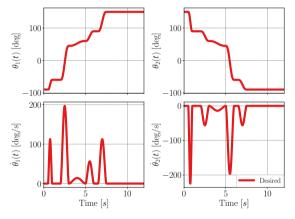


Figure 7: Desired angle position and rate.

Control Synthesis Revisited

- The subcontrollers to be gain scheduled will be SPR controllers with feedthrough, which are in turn VSP.
- The KYP lemma and gain matrix K from the LQR problem are used to synthesize the SPR controllers [Benhabib et al., 1981].
- ► The LQR problem requires the linearized system dynamics.
- Since the SPR controller is a rate-based controller, a proportional control prewrap is then added to the system to control the joint displacements of the system.
 - This prewrap does not violate the passive map of the system [Márquez, 2003].

Control Synthesis

The linearization of the prewrapped model about q

i is given by

$$\delta \dot{\mathbf{x}}(t) = \mathbf{A}_i \delta \mathbf{x}(t) + \mathbf{B}_i \delta \mathbf{u}(t), \quad \delta \mathbf{y}(t) = \mathbf{C}_i \delta \mathbf{x}(t),$$

with

$$\mathbf{A}_i = egin{bmatrix} \mathbf{0} & \mathbf{1} \ -ar{\mathbf{M}}^{-1}(ar{\mathbf{q}}_i)\mathbf{K}_\mathrm{p} & \mathbf{0} \end{bmatrix}, \ \mathbf{B}_i = egin{bmatrix} \mathbf{0} \ ar{\mathbf{M}}^{-1}(ar{\mathbf{q}}_i) \end{bmatrix}, \ \mathbf{C}_i = egin{bmatrix} \mathbf{0} \ \mathbf{1} \end{bmatrix}^{\mathsf{T}},$$

where $\bar{\mathbf{M}}(\bar{\mathbf{q}}_i)$ is the **measured mass matrix** and \mathbf{K}_{p} is the proportional gain matrix.

The gain matrix K_i is computed for each linearization point by solving the algebraic Riccati equation (ARE).

Table 1: Two-Link Manipulator Properties

Link Parameters	Link 1	Link 2
Length [m]	$L_1 = 1.10$	$L_2 = 0.85$
Measured Length [m]	$\bar{L}_1 = 1.08$	$\bar{L}_2 = 0.83$
Mass [kg]	$m_1 = 0.40$	$m_2 = 0.90$
Measured Mass [kg]	$\bar{m}_1 = 0.44$	$\bar{m}_2 = 0.99$

Table 2: Controller Design Parameters

Properties	Symbol	Value
Proportional Gain	\mathbf{K}_{p}	$\operatorname{diag}(35, 35)$
LQR Weights	$egin{array}{c} Q_{ m LQR} \ R_{ m LQR} \end{array}$	$\frac{\text{diag}(0.33, 0.25, 180, 180)^{-2}}{\text{diag}(15, 15)^{-2}}$
Feedthrough	δ	0.0001

Control Synthesis

SPR control synthesis is then completed by using the KYP lemma to set

$$\mathbf{A}_{\mathrm{c},i} = \mathbf{A}_i - \mathbf{B}_i \mathbf{K}_i, \qquad \qquad \mathbf{B}_{\mathrm{c},i} = \mathbf{P}_i^{-1} \mathbf{K}_i^{\mathsf{T}}, \qquad \qquad \mathbf{C}_{\mathrm{c},i} = \mathbf{K}_i,$$

where $\mathbf{P}_i = \mathbf{P}_i^{\mathsf{T}} \succ 0$ is the solution to the Lyapunov equation, $\mathbf{A}_{c,i}^{\mathsf{T}} \mathbf{P}_i + \mathbf{P}_i \mathbf{A}_{c,i} = -\mathbf{Q}_i$, for $\mathbf{Q}_i = \mathbf{Q}_i^{\mathsf{T}} \succ 0$ [Benhabib et al., 1981].

- To make the SPR controller VSP, an arbitrary small feedthrough term $D_c = \delta 1$ is added.
- ► Therefore, for each linearization point q
 _i, a VSP controller, G_i: L_{2e} → L_{2e}, can be synthesized with the state-space form

$$\dot{\mathbf{x}}_i(t) = \mathbf{A}_{\mathrm{c},i} \mathbf{x}_i(t) + \mathbf{B}_{\mathrm{c},i} \mathbf{u}_i(t), \qquad \qquad \mathbf{y}_i(t) = \mathbf{C}_{\mathrm{c},i} \mathbf{x}_i(t) + \mathbf{D}_{\mathrm{c}} \mathbf{u}_i(t).$$

Scheduling Signals

For the three linearization points $\bar{q}_1,\,\bar{q}_2,$ and $\bar{q}_3,$ the scheduling signals in Figure 8 are defined as

$$s_{1}(t) = \begin{cases} 1 - \left(\frac{t}{3}\right)^{4} & 0.0 \le t \le 3.0, \\ 0 & 3.0 < t, \end{cases}$$
(3a)
$$s_{2}(t) = \begin{cases} 1 - \left(\frac{t-3}{2.8}\right)^{4} & 0.2 \le t \le 5.8, \\ 0 & \text{otherwise,} \end{cases}$$
(3b)
$$s_{3}(t) = \begin{cases} 0 & 0.0 \le t < 5.0, \\ 1 - \left(\frac{t-7.5}{2.5}\right)^{4} & 5.0 \le t \le 7.0, \\ 1 & 7.0 < t. \end{cases}$$
(3c)

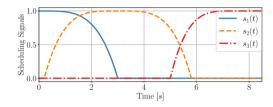


Figure 8: Scalar scheduling signals $s_1(t)$, $s_2(t)$, and $s_3(t)$ defined in (3).

Remark

As required by [Damaren, 1996] and [Forbes, Damaren, 2010], all scheduling signals are bounded, and at all times, at least one scheduling signal is active.

Scheduling Matrices

For $\mathbf{u} : \mathbb{R}_{\geq 0} \to \mathbb{R}^2$, the scheduling of each subcontroller \mathcal{G}_i requires five hyperparameters: • one α_i ,

• four scheduling signals for the scheduling matrix $\mathbf{\Phi}_i \in \mathbb{R}^{2 \times 2}$.

Using three subcontrollers, one such set of scheduling matrices are

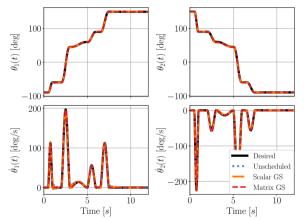
$$\begin{split} \Phi_{1}(t) &= \begin{bmatrix} 2s_{1}(t) + 4s_{2}(t) & 0\\ 0 & s_{1}(t) \end{bmatrix}, & \alpha_{1} = 2, \end{split} \tag{4a} \\ \Phi_{2}(t) &= \begin{bmatrix} s_{2}(t) & 0\\ s_{2}(t) & s_{2}(t) \end{bmatrix}, & \alpha_{2} = 1, \end{aligned} \tag{4b} \\ \Phi_{3}(t) &= \begin{bmatrix} s_{3}(t) + 2s_{2}(t) & 0\\ 0 & s_{3}(t) \end{bmatrix}, & \alpha_{3} = 2, \end{aligned} \tag{4c}$$

where $s_1(t)$, $s_2(t)$, and $s_3(t)$ are defined in (3).

Results

Three different control approaches:

- 1. **Unscheduled**: A single VSP controller designed about the linearization of the robot at the end of its trajectory.
- Scalar GS: Three VSP subcontrollers scheduled as per [Forbes, Damaren, 2010]
- 3. **Matrix GS**: Three VSP subcontrollers scheduled using proposed matrix-gain-scheduling architecture.



Remark

Across all three control approaches, the exact same $K_{\rm p}, Q_{\rm LQR}, R_{\rm LQR}$, and δ are used for the synthesis of the VSP subcontrollers.

Results

Table 3: RMS Error of Joint Angle and Joint Angle Rate

	RMS angle error [deg]		RMS angle rate error [deg/s]	
Control method	e_1	e_2	\dot{e}_1	\dot{e}_2
Unscheduled	0.8328	0.6688	2.5933	1.5587
Scalar scheduling	0.6839	0.6464	2.1307	1.2702
Matrix scheduling	0.0668	0.4515	0.1480	1.1352

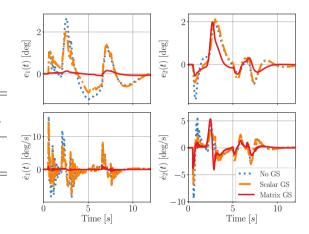
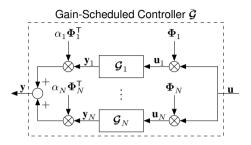


Figure 9: Comparison of joint angles errors and error rates.

Summary



Proposed a novel gain-scheduling architecture using scheduling matrices.

- The gain-scheduled controller is
 - ▶ **ISP** if each *G_i* is ISP and the scheduling matrices are **strongly active**.
 - **OSP** if each \mathcal{G}_i is OSP and the scheduling matrices are **active**.
 - ▶ VSP if each *G*_i is VSP and the scheduling matrices are strongly active.
- The conditions on the scheduling matrices can be interpreted as an extension of the conditions on the scheduling signals reported in [Damaren, 1996] and [Forbes, Damaren, 2010].

Ongoing and Future Work

The gain-scheduling of a more general class of passive systems is considered in:

- Conic Gain-Scheduled Control of an Aeroelastic Airfoil [Caverly, Brown, 2021],
- Gain-Scheduled QSR-Dissipative Systems: An Input-Output Approach [Anderson, Caverly, Lamperski, 2023]. These papers also use the same scalar-gain-scheduling architecture presented in [Damaren, 1996].
- In our TAC paper under review, the proposed matrix-gain-scheduling architecture can be extended to consider the gain-scheduling of QSR-dissipative systems for
 - Case 1: All the N subsystems are QSR-dissipative with $\mathbf{Q}_i = \mathbf{Q}_i^{\mathsf{T}} \prec 0$.
 - ► Case 2: All the *N* subsystems are QSR-dissipative with either $\mathbf{Q}_i = \mathbf{Q}_i^{\mathsf{T}} \prec 0$ or $\mathbf{Q}_i = \mathbf{Q}_i^{\mathsf{T}} \preceq 0$, and share a common $\mathbf{S}_i = \mathbf{S} \in \mathbb{R}^{n_y \times n_u}$.
- How to optimally design the scheduling matrices?

• Can α_i be time varying?

Questions?

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Paper/Slides/Code



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Backup Slides: Control Effort

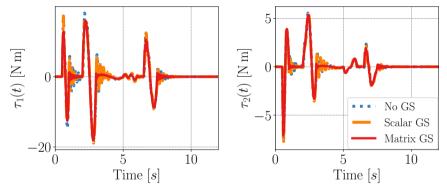


Figure 10: Comparison of joint torques.

Backup Slides: Negative Feedback Interconnection

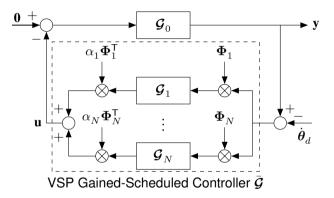


Figure 11: Gain-scheduled feedback control of the plant to be controlled \mathcal{G}_0 , prewrapped with proportional control, and the gain-scheduled controller $\overline{\mathcal{G}}$.

Backup Slides: Trajectory Generation

Trajectory generation is achieved similar to [Forbes, Damaren, 2010], by choosing discrete joint angles $\theta_d(t_k)$ and $\theta_d(t_{k+1})$ at times t_k and t_{k+1} , and interpolating between them as such

$$\eta(t) = \frac{t - t_k}{t_{k+1} - t_k},\tag{5a}$$

$$p_5(t) = 6\eta^5 - 15\eta^4 + 10\eta^3, \tag{5b}$$

$$\boldsymbol{\theta}_{\mathrm{d}}(t) = p_{5}(t) \left(\boldsymbol{\theta}_{\mathrm{d}}(t_{k+1}) - \boldsymbol{\theta}_{\mathrm{d}}(t_{k})\right) + \boldsymbol{\theta}_{\mathrm{d}}(t_{k}).$$
(5c)

The desired discrete joint angles are chosen such that the joint angles operate within $[-90^{\circ}, 150^{\circ}]$.

Backup Slides: Expanding OSP Coefficient for base case

The special case of $\alpha_i = 1$ and $\Phi_i(t) = s_i(t)\mathbf{1}$ is referred to as the *base case*. As discussed in our TAC paper under review, for the base case, the gain-scheduled OSP coefficient $\bar{\varepsilon}$ can be expanded as

$$\bar{\varepsilon} = \frac{\varepsilon_{\min}}{\alpha_{\max}^2 \bar{\sigma}_{\Psi}^2} = \frac{\varepsilon_{\min}}{\sup_{t \in [0,T]} \sum_{i \in \mathcal{N}} |s_i(t)|^2} > 0.$$
(6)

Backup Slides: Finite \mathcal{L}_2 Gain

Using the Cauchy–Schwartz inequality, it can be shown that if a system is OSP with $\mathbf{Q} = -\varepsilon \mathbf{1}$, it also possesses finite \mathcal{L}_2 gain such that $\varepsilon = 1/\gamma$. Here, by defining $\varepsilon_i = 1/\gamma_i$ with $\gamma_i \in \mathbb{R}_{>0}$, (6) can be rewritten as $\varepsilon = 1/\gamma$, where

$$\gamma = \max_{i \in \mathcal{N}} (\gamma_i) \sup_{t \in [0,T]} \sum_{i \in \mathcal{N}} |s_i(t)|^2 > 0.$$

$$\tag{7}$$

Therefore, the gain-scheduled system $\overline{\mathcal{G}}$ posses finite \mathcal{L}_2 gain with γ defined as per (7). In [Forbes, Damaren, 2010], it is shown that the gain-scheduled controller has finite \mathcal{L}_2 gain given by

$$\gamma = \sum_{i \in \mathcal{N}} \|s_i\|_{\infty}^2 \gamma_i > 0.$$
(8)